

LP3350

LED Driver with PWM Dimming Control

General Description

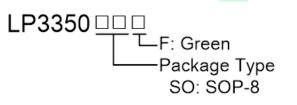
LP3350 is a cost effective LED driver optimized for LCD monitor and LCD TV backlighting application. It provides a high performance LED backlight solution with minimized bill of material count.

The LP3350 contains a PWM boost driver which uses current mode control and fixed frequency operation to regulate the LED current. The LED current is sensed through an external current sense resistor. The voltage across the sensing resistor is compared with reference level of 0.3V, the error amplified to control the pulse width of the power switch thus to regulate the current flowing the LED.

Otherwise, The LP3350 offers external frequency PWM dimming method for a wide range of dimming control.

Other features include over current protection (OCP), output over voltage protection (OVP), and under-voltage lockout (UVLO). The LP3350 is available in a space saving SOP-8 package.

Order Information



Features

- ♦ Wide V_{IN} Range: 8V to 28V
- Current-Mode PWM Controller
- External PWM Dimming Mode
- Under-Voltage Lockout
- Over Voltage Protection
- Over Current Protection
- Under-Voltage Protection
- Over-Temperature Protection
- Available in SOP-8
- RoHS Compliant and Halogen Free
- Pb-Free Package

Applications

- ♦ TFT LCD TV
- TFT LCD Monitor
- Flat Panel Display

Marking Information

w D c	Device	Marking	Package	Shipping	
	LP3350	LPS	SOP-8	4K/REEL	
话	坐道 髀	LP3350			
1123		YWX			
	Y is year code. W is week code. X is series number.				



Typical Application Circuit

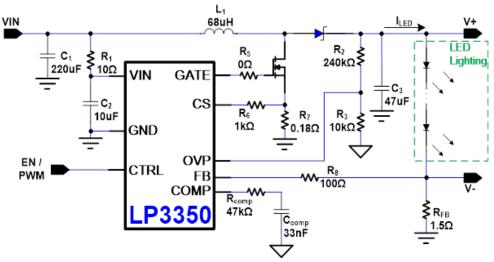


Figure 1. Typical Application Circuit of LP3350

Pin Configuration

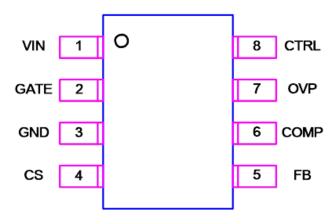


Figure 2. SOP-8 Package Top View



Function Block Diagram

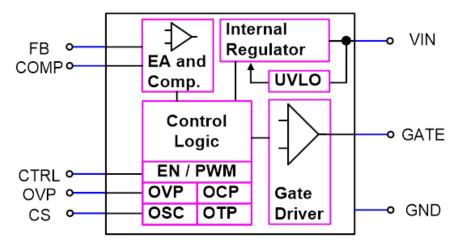


Figure 3. Function Block Diagram

Functional Pin Description

Pin NO.	Pin Name	Description		
1	VIN	Input Supply Pin. Decouple with 10µF ceramic capacitor close to the pin.		
2	GATE	External NMOS Gate Drive Pin.		
3	GND	Ground. IOWPOWEPSemi		
4	CS	Current Sense Input Pin.		
5	FB	Regulator Feedback Input. Connect to an external resistive to set the output current.		
6	6 COMP Regulator Error Amplifier Compensation Pin.			
7	OVP	Over Voltage Protection Sense Input. Connect to an external resistive voltage divider from the V+ to GND.		
8	CTRL	Enable and External PWM Dimming Control.		



Absolute Maximum Ratings Note 1

\prec	VIN to GND0.3V to +36V			
\prec	GATE, CS, FB, COMP, OVP, CTRL to GND			
\prec	Operating Junction Temperature Range (T _J)			
\prec	Operation Ambient Temperature Range			
\prec	Storage Temperature Range			
\prec	Maximum Soldering Temperature (at leads, 10sec)+260°C			
\prec	Maximum Junction Temperature+150°C			
Note 1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress				
r	tings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of			

ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Thermal Information

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Thermal Resistance			
SOP-8, θ _{JA}			 112°C/W
SOP-8, θ _{JC}	L P Semi	LowPowerSemi 微源半導體	 39°C/W





Electrical Characteristics

(V_{IN}=12V, V_{CTRL}=5V, T_A=25°C, unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Units	
General							
Input Supply Voltage	V _{VLI}		8		28	V	
N/ Oursely Ourseat		Sleep Current		310	400	uA	
V _{IN} Supply Current	Ι _Q	Operation Current		2.5	3	mA	
Input UVLO Threshold	V _{UVLO(VTH)}	V _{IN} Rising	6	7	8	V	
UVLO Threshold Hysteresis	V _{UVLO(HYS)}	Falling Hysteresis		500		mV	
Soft Start Slope	T _{SS_Slop}			400		mV/ms	
Thermal Shutdown Threshold	T _{SD}	Temperature Rising	140	160	180	°C	
Thermal Shutdown Hysteresis	ΔT_{SD}			30		°C	
PWM Control							
Internal Oscillator Frequency	Fosc		175	200	225	kHz	
Maximum Duty Cycle	D _{MAX}		93	95	97	%	
CTRL Threshold Voltage	V _{IH}	Logic High.			2.6	N	
	Vileni	Logic Low	1			V	
Pull Down Resistor	R _{Pull_Low}	LowPowerSemi		150		kΩ	
Error Amplifier		治 酒半道體					
Reference Voltage	V _{REF}	Reference voltage at non-inverting input.	0.289	0.3	0.311	V	
Open Loop Voltage Gain	Am			70		dB	
Transconductance of EA	Gm		80	100	120	uA/V	
GATE Source Current	I _{Source}		24	30	36	uA	
GATE Sink Current	I _{Sink}		48	60	72	uA	
Protection Threshold							
Over Voltage Protection	V _{OVP}	Threshold of OVP	1.8	2	2.2	V	
Over Current Protection	V _{OCP}	Threshold of OCP		330		mV	
Fault Trigger Duration	T _{Fault}			50		ms	
Connect detect time	T _{Dly}			1		ms	
SCD Throshold		Normal Operation		200		mV	
SCP Threshold	V _{OVP_UV}	System Startup		100		mV	



Application Information

The LP3350 is designed in a current mode, constant frequency PWM boost converter. It can use dimming input that can by external control signal with a duty ratio of 1%-100% in 100Hz to 1kHz. LP3350 offers protection features to protect the system such as output over voltage protection, boost diode disconnection protection, output short circuit protection and over temperature protection.

Under Voltage Lockout (UVLO)

The LP3350 had an UVLO internal circuit that enable the device once the voltage on the VIN voltage exceeds the UVLO threshold voltage.

Boost Controller

The LP3350 uses 200kHz fixed-frequency, current mode architecture to fixed the output current. The output voltage automatically adjusts its voltage to the LED forward voltage to improve performance.

Boost Loop Compensation

The feedback loop can be compensated with an external compensation network consisted of Rcomp, Ccomp (As Figure 1). Choosing Rcomp to set high frequency integrator gain for fast transient response and Ccomp to set the integrator zero to maintain loop stability.

Over Voltage Protection

The LP3350 converter has an over voltage protection by OVP pin. When the LEDs fail open circuit or LEDs are disconnected from the circuit, the over voltage function will monitor the output voltage through OVP pin to protect the converter. When LP3350 occur OVP, it will latch off until VIN is re-startup or CTRL input is recycled.

LED Current Setting

The LED current is specified by current sense resistor between the FB pin to ground. In order to have accurate LED current, precision resistors are preferred. The LED current can be programmed by:

 $I_{LED}=0.3/R_{FB}$

Dimming Control

The LED brightness is controlled by the PWM signal at CTRL pin which has different duty cycle. LP3350 can accept an external PWM signal to CTRL pin in the range of 100Hz to 1 kHz.

Over Temperature Protection

The LP3350 device enters over temperature protection(OTP) if its junction temperature exceeds 160°C (Typ.). During over temperature protection none of the device's functions are available. To resume normal operation the junction temperature need cool down, and the outputs will restart.

Current Sense and Over Current Protection (OCP)

The over current protection level can be set by R7(As Figure 1) and senses inductor current to compare with current limit value. When the inductor current exceeds the current limit, the switching will turns off immediately. It prevents large current damaging the external component. The OCP level is calculated as:

I_{OCP}=0.33/R₇

The LP3350 uses a current mode control structure. The CS pin not only has current sense function but also build-in a slope compensation to avoid sub-harmonic oscillation. Additional slope compensation lessened the influence of the sensed current in the control loop.

Layout Guideline

The proper PCB layout and component placement are critical for all circuit. The careful attention should be prevent electromagnetic interference (EMI) problems. Here are some suggestions to the layout of LP3350 design.

1. Connected all ground together with one uninterrupted ground plane with at least two vias .

2. The input capacitor should be located as closed as possible to the VIN and ground plane.

3. Minimize the distance of all traces connected to the LX node, that the traces short and wide route to obtain optimum efficiency.

4. All output capacitor must be closed to ground plane. The ground terminal of COUT must be located as closed as possible to ground plane.

5. Radiated noise can be decreased by choosing a shielded inductor.

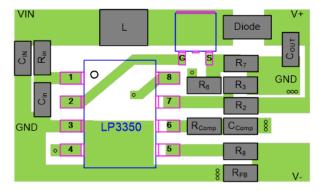
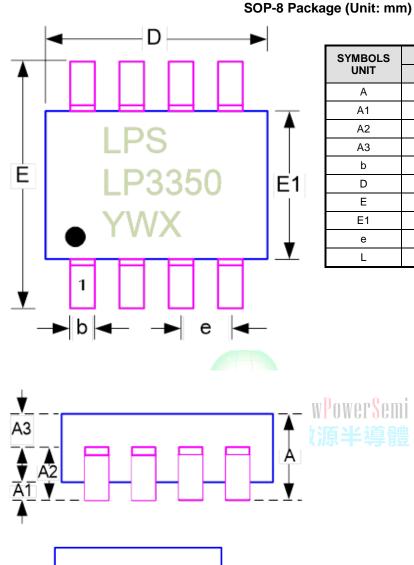


Figure 4. Recommended PCB Layout Diagram





Outline Information



DIMENSION IN MILLIMETER SYMBOLS UNIT NOM MIN MAX ----1.750 А A1 0.100 --0.225 A2 1.300 1.400 1.500 A3 0.600 0.650 0.700 b 0.390 ---0.470 D 4.800 4.900 5.000 Е 5.800 6.200 6.000 E1 3.900 3.800 4.000 1.27BSC е L 0.500 ---0.800